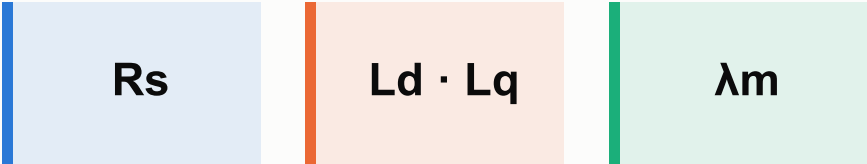


How an inverter measures the motor it drives

Automatic identification of R_s , L_d , L_q and λ_m with the hardware already on the board



R_s

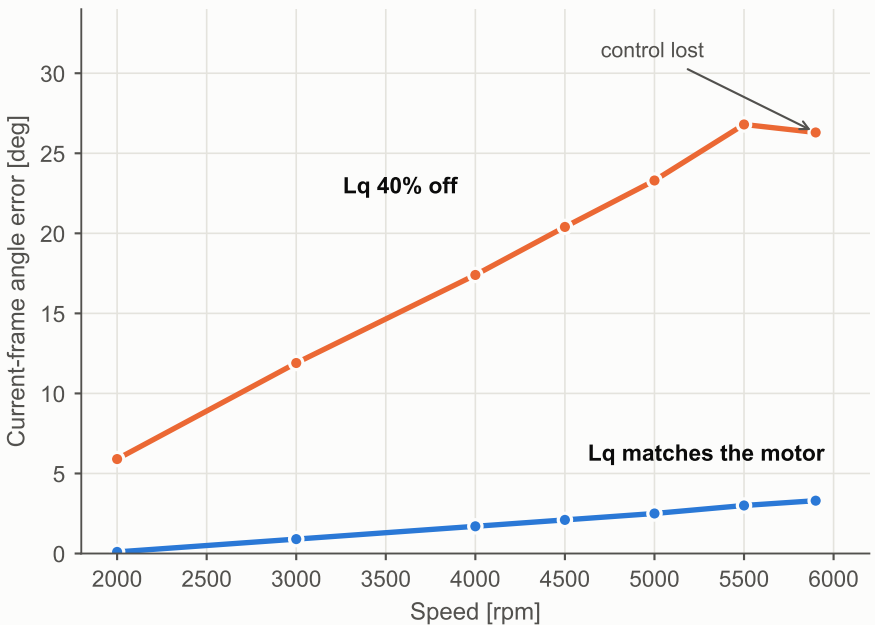
$L_d \cdot L_q$

λ_m

THE PROBLEM

The observer is only as good as its motor parameters

A sensorless DTFC estimates the rotor angle from a motor model. If L_q is off, the angle error grows with current and speed, until the control is lost.



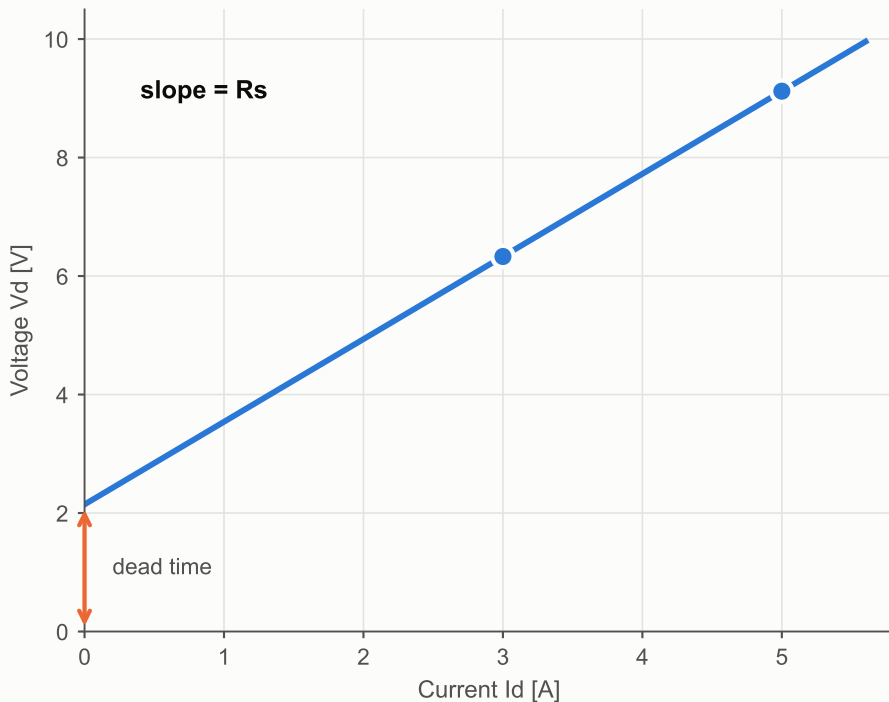
Stator resistance R_s

Two DC current levels during rotor alignment.

The slope of V vs. I is R_s .

The intercept is the inverter dead-time drop:

it cancels out of R_s and is reused in step 3.



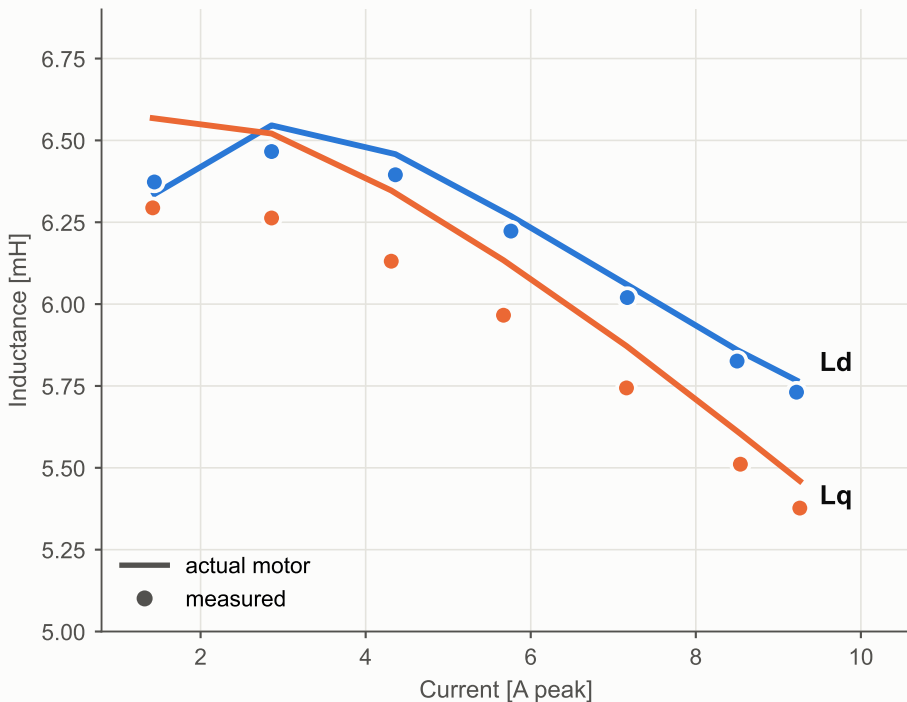
Inductances $L_d(I_d)$, $L_q(I_q)$

HF sinusoidal voltage injected on d, then q;

synchronous demodulation solves $v = R \cdot i + L \cdot di/dt$.

Sweeping the current gives the saturation curves.

L_q uses AC current: zero mean torque, no rotation.



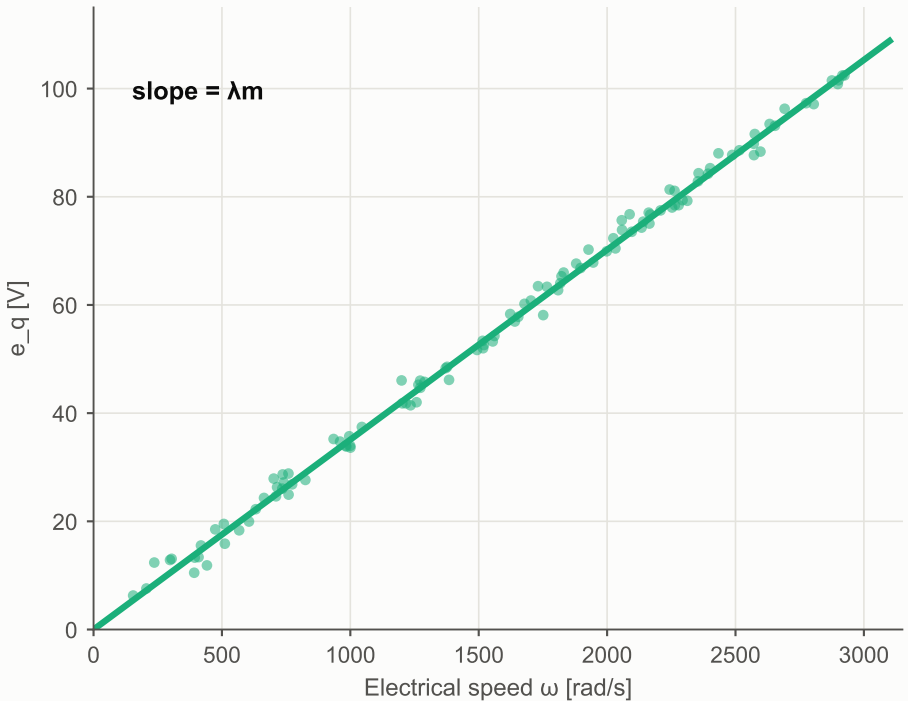
Magnet flux linkage λ_m

q-axis back-EMF after removing the R, L and

dead-time drops: $e_q = \lambda_m \cdot \omega$.

Least squares with forgetting gives λ_m ,

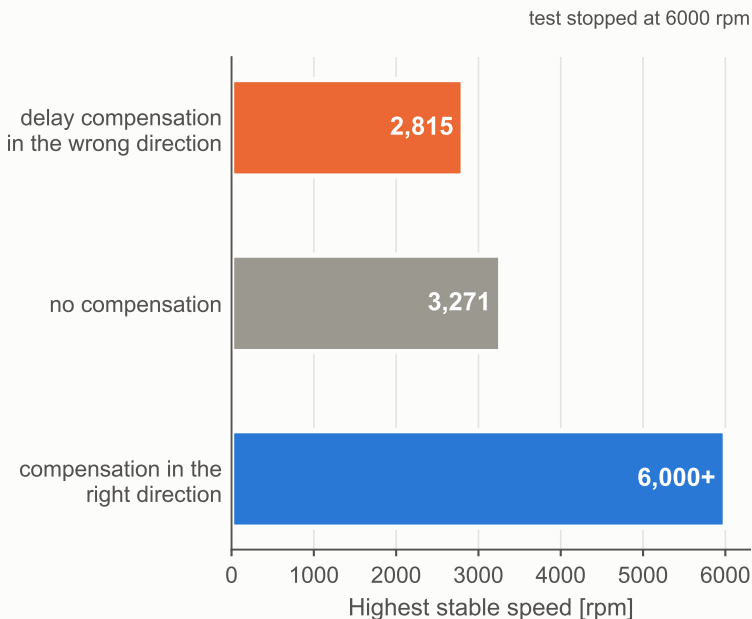
which tracks the magnet temperature.



LESSON LEARNED

Measuring is easy. Timing is hard.

Between sampling the current and applying the voltage there are ~1.5 PWM periods. At 6000 rpm that is 34° of rotation. Compensating it in the right direction moved the stability limit:



RESULTS

What the inverter now knows on its own

Rs	±0.4%	standstill, every start
Ld, Lq	±4%	standstill, full saturation curves
λm	±3%	online, tracks temperature
Dead time	±0.3%	standstill, reused online

Validated in simulation with the real firmware.

How do you identify motor parameters in your drives?